

17-4-55

# Beat Oscillator Coil

Winding # 31 B+S .0089" dia

Winding  $2\frac{3}{8}$ " long,  $7\frac{1}{2}$  turns

Coil form 3" long,  $1\frac{1}{8}$ " dia.

Holes for #8 BA screws  $\frac{1}{8}$ " from edge

Inductance 732  $\mu$ h at 790 kc

Freq C Q

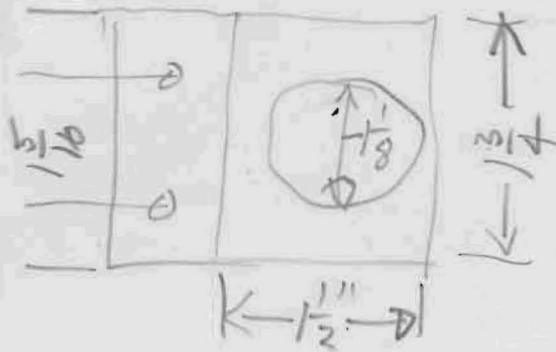
280 kc 463 pf 89

~~560 kc~~ 1125 pf 120 >

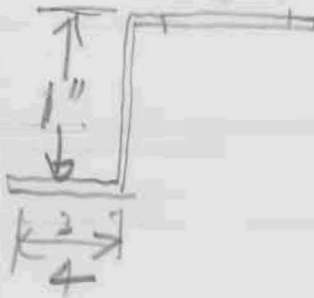
1100 kc 27 pf 116

Primary  $1\frac{3}{32}$ " long.

# Bracket for socket



pc  $3 \frac{1}{4} \times 1 \frac{3}{4}$



Dr. to

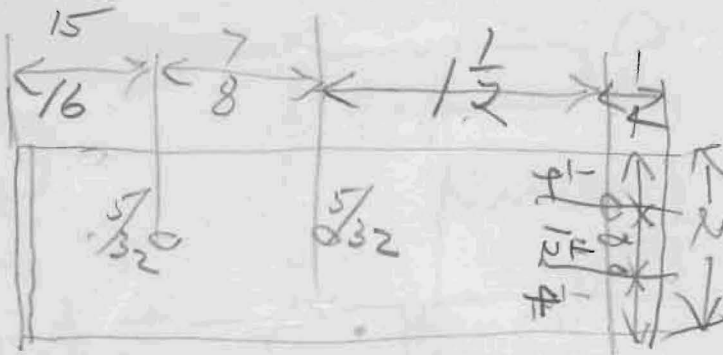
$\frac{1}{4}$   
 $\frac{15}{16}$   
 $\frac{3}{16} \times 2$   
 $\frac{5}{16} \times 2$   
 $\frac{4}{16} \times 2$   
 $\frac{11}{16}$   
 $\frac{11}{16}$   
 $\frac{11}{16}$

DATE,

19

AMOUNT.

£ s. d.



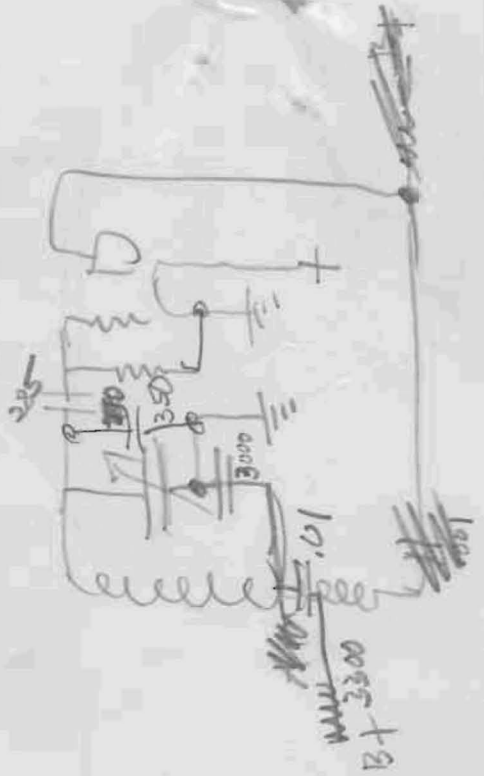
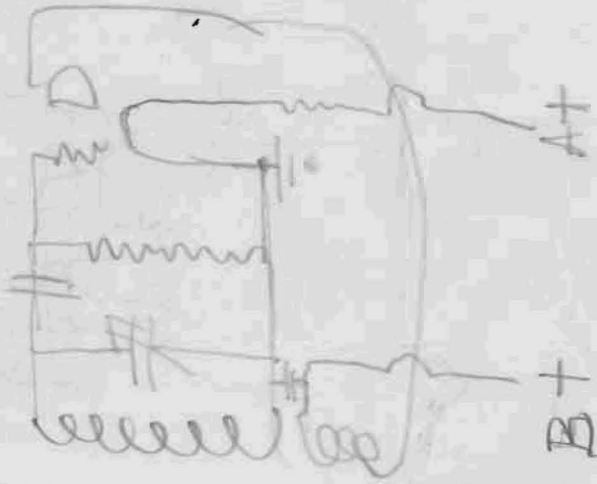
*Bracket for condenser*

$$\frac{420}{165} \text{ variation}$$

$$\frac{655}{75} \text{ variation}$$

$$\frac{580}{580} \text{ variation}$$

550pF  
short





## 2nd I.F. Transformer

Primary = 3" I.D.,  $\frac{3}{4}$ " long, 4 Pi

D.C. resistance: Primary 61 ohms, Secondary 131 ohms

1 KC inductance: Pri. 16.7 mh, Sec. 95.0 mh,

$L_p + L_s + 2M = 136$  mh,  $L_p + L_s - 2M = 84$  mh,  $M = 13$  mh,  $K = 33\%$

## Radio Frequency Measurements on Q machine,

Assembled Transformer with diodes connected,

	Freq KC	Cap pf	Q	R ohms		
Primary	60	390	85	76	$C_0 =$	23.3 pf
	120	80	85	151	$L_0 =$	17.0 mh
	173	27?	40	462	$R =$	107 KC/Q
Secondary	50	98	167	180	$C_0 =$	8.0 pf
	70.7	45	155	275	$L_0 =$	96 mh
	89	27?	130	412	$R =$	602 KC/Q
$L_p + L_s + 2M$	50	63	235	186	$C_0 =$	10.0 pf
	70.7	26.5	215	287	$L_0 =$	139 mh
$L_p + L_s - 2M$	50	108	160	166	$C_0 =$	12.0 pf
	70.7	48	165	218	$L_0 =$	85 mh
	89	27?	155	305	$R =$	531 KC/Q

Q probably is too high causing R to be too low  
 $L_0$  is correct.

when  $\omega_2 = 2\omega_1$ ;  $C_0 = (C_1 - 4C_2)/3$

$M = 13.5$  mh,  $K = 33\%$

when  $\omega_2 = \sqrt{2}\omega_1$ ;  $C_0 = (C_1 - 2C_2)/1$

Capacity for 250 KC

Primary 24.3 pf

Secondary 4.27 pf

when  $\omega_2 = 3\omega_1$ ;  $C_0 = (C_1 - 9C_2)/8$



150

200

250

350

400

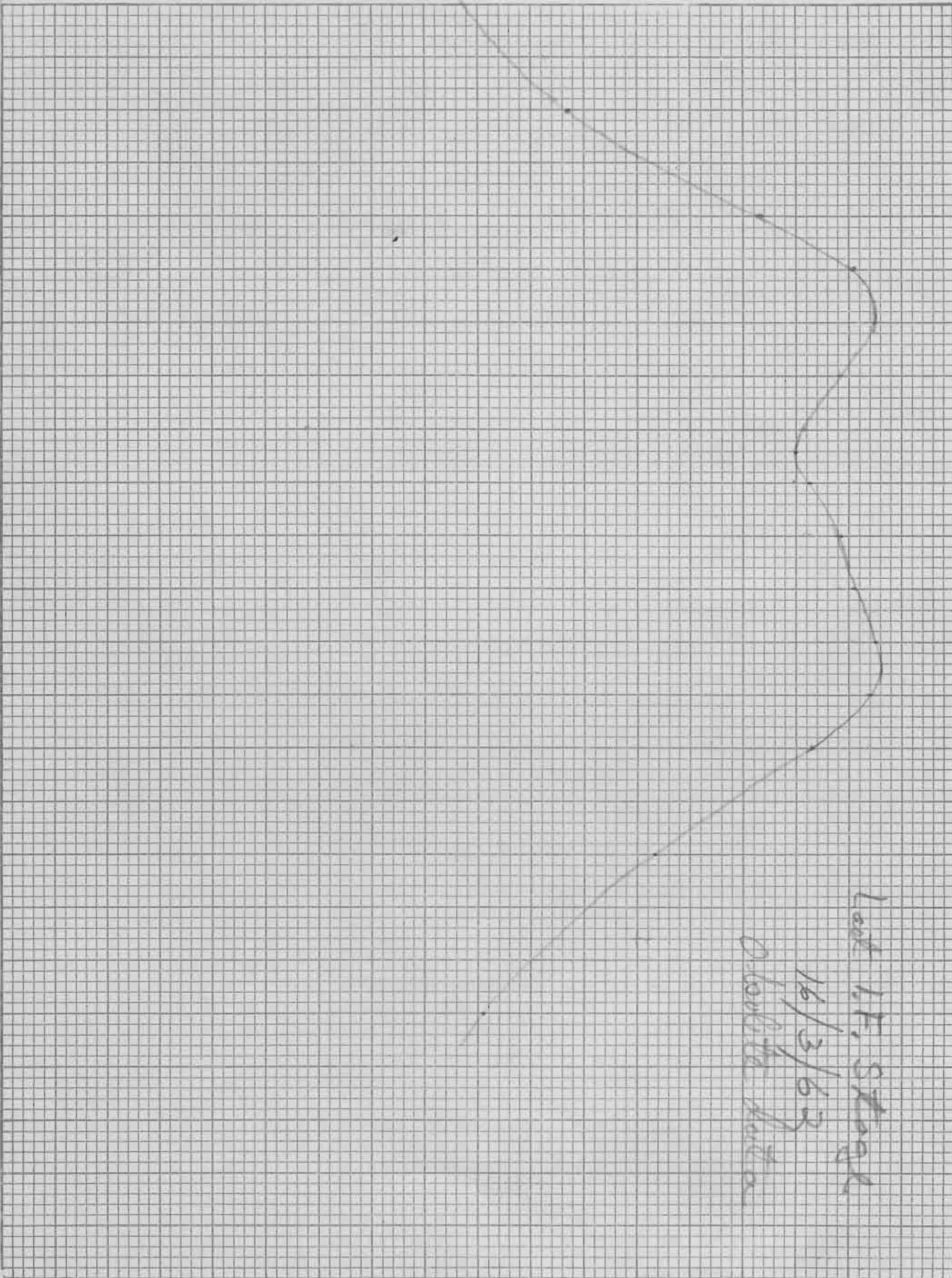
18 17 16 15 14 13 12 11 10 9 8 7 6 5 4 3 2 1

10

20

30

DB



Last 1/2 Stack  
 16/3/67  
 Double data

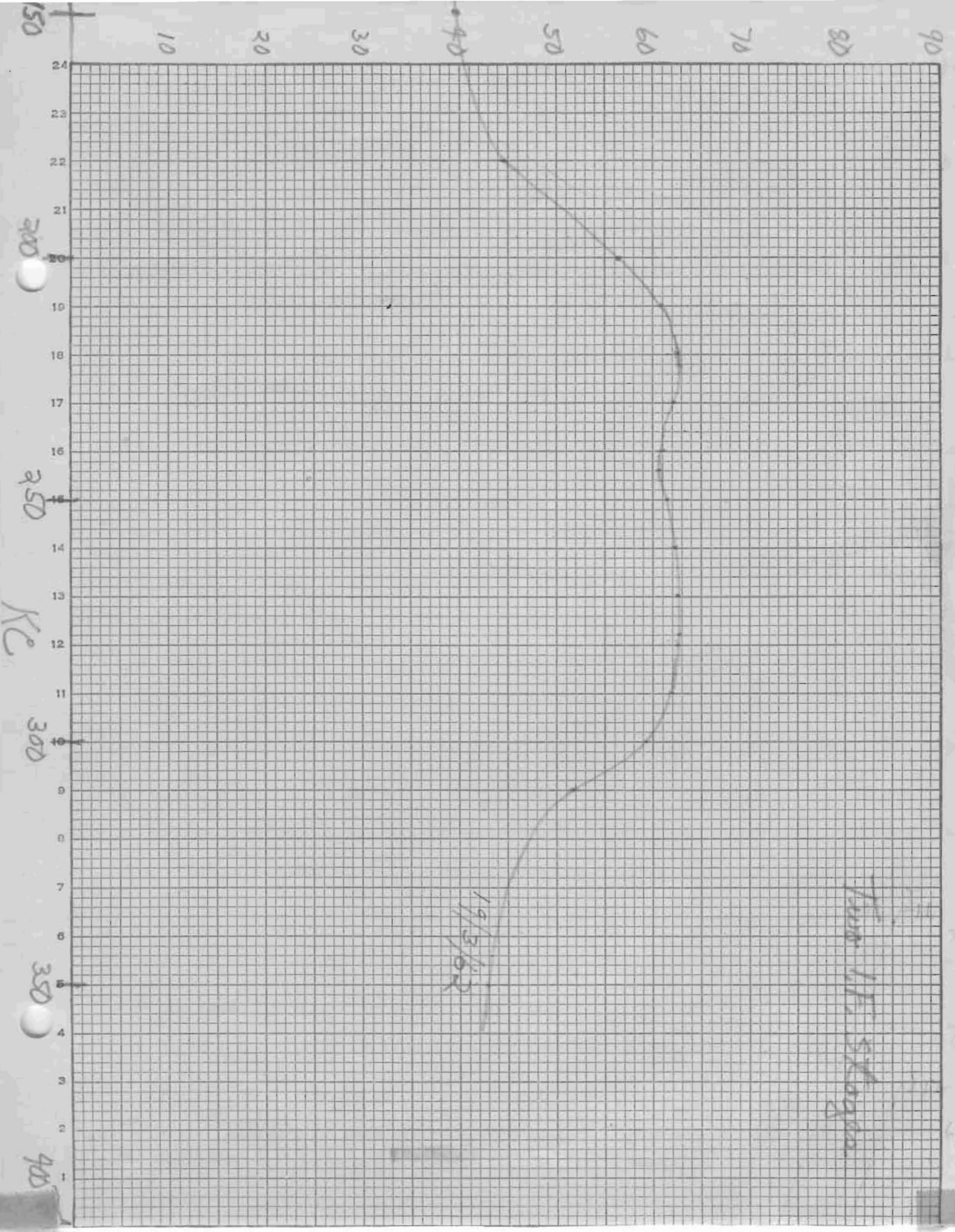
19/3/62

## Comparison of Voltage Indications on General Radio Signal Generator with Hewlett Packard Vacuum Tube Voltmeter.

Signal Generator	VTVM 1 volt scale	Signal Generator	VTVM 3 volt scale
0.2	0.21	0.8	0.85
0.36	0.36	1.4	1.41
0.80	0.80	2.0	2.08
		3.0	3.10

The VTVM is a peak reading instrument calibrated in rms for a sine wave. Actual peak voltage at input is always  $\sqrt{2} = 1.414$  times the meter indication irrespective of wave shape.





150

10

DB

20

30

200

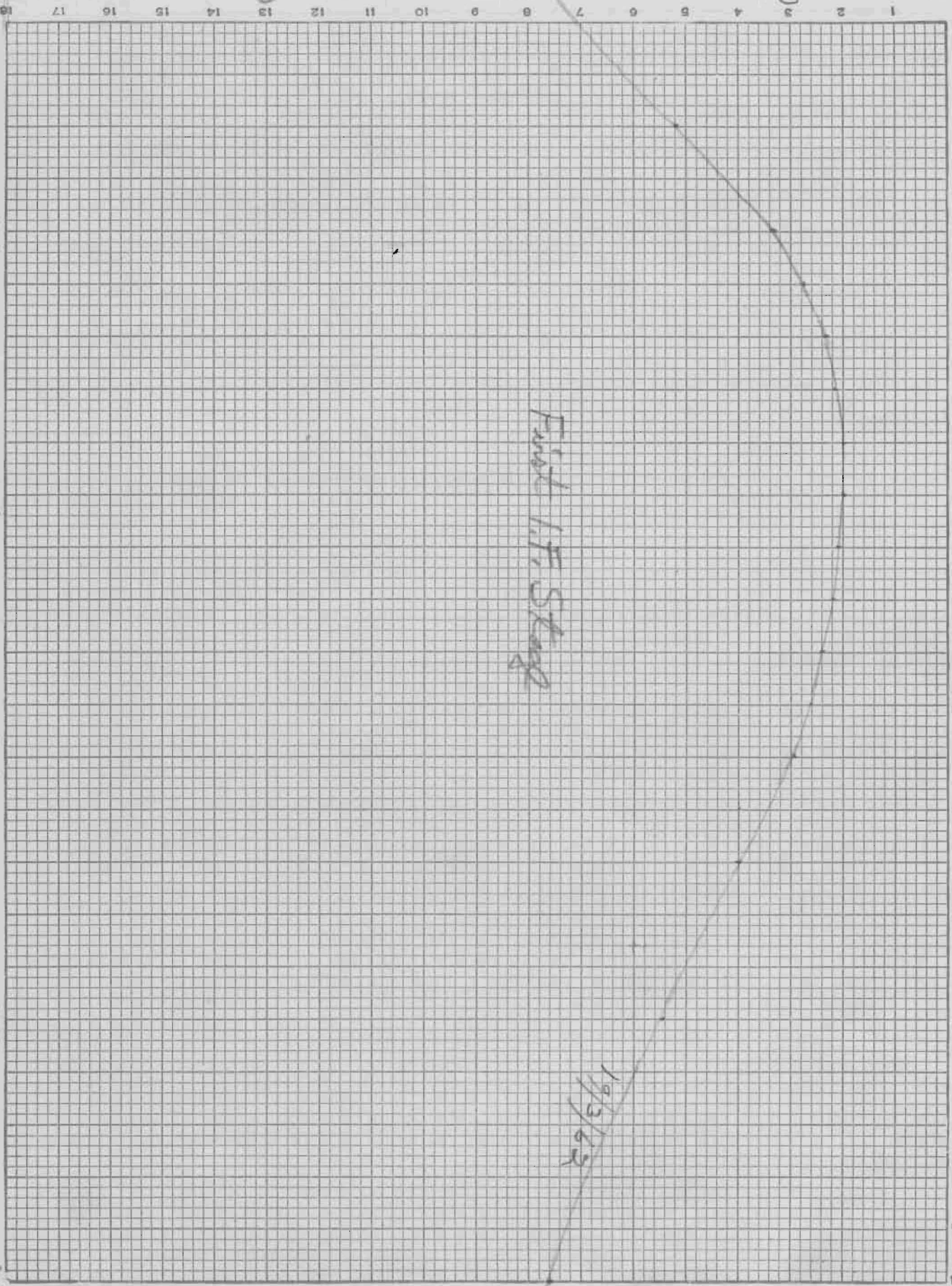
250

KE

300

350

400



First I.F. Stage

1/3/62

19/3/62

Obsolete

First I.F. Stage, 0.5V output, 51.8ma tube  
 15K $\Omega$  cathode resistor, 56K $\Omega$  load, H.P. VTVM.

Gain vs Frequency  
 $E_c = +12V$

Gain versus Load Resistance

Freq. KC	Input mv	Gain DB	Freq. KC	Input mv	Gain DB	
150	50	20.0	204	9.4	34.5 Peak	
180	26	25.7	242	11.4	32.9 Dip	
200	17.1	29.4	294	9.6	34.4 Peak	
210	15.0	30.5			220K $\Omega$ load	
220	13.5	31.4	183	6.4	37.9 Peak	
230	13.0	31.7	220	10.5	33.6 Dip	
240	12.6	32.0	280	5.2	39.7 Peak	
250	12.6	32.0			Load removed	
260	12.8	31.9	180	3.3	43.6 Peak	
270	13.0	31.7	223	10.6	33.5 Dip	
280	13.6	31.3	284	2.1	47.6 Peak	
290	14.5	30.8				
300	15.7	30.1			Overload and Clipping	
320	20.0	28.0	Input	Output	Input	Output
350	28	25.1	mv	volts	mv	volts
400	46	20.8	1	0.04	100	1.31
			2	0.08	200	1.39
			5	0.20	360	1.41
			10	0.40	800	1.42
			20	0.77	1400	1.47
			50	1.24	2000	1.51
				(over)	3600	1.53

Regulation at  $E_c = +12V$   
 $E_b$  volts 157 166 175  
 $I_k$  ma 0.99 1.01 1.03  
 Output %  $\leq 0.1$  100  $\leq +0.1$   
 imperceptible change



The fixed cores in primary & secondary are in contact without airgap to give tightest coupling.

### First I.F. Transformer

Capacity coupling aids inductive coupling.

D.C. resistance: Primary 49 ohms, Secondary 49 ohms

1 KC inductance: Pri. 25.9 mh, Sec 24.4 mh

$L_p + L_s + 2M = 60.7 \text{ mh}$ ,  $L_p + L_s - 2M = 39.5 \text{ mh}$ ,  $M = 5.3 \text{ mh}$ ,  $K = 21\%$

### Radio Frequency Measurements on G machine

	Freq KC	Cap pf	Q	R ohms		
Primary	50	381	140	59	$C_0 = 5.7 \text{ pf}$	
	100	91	164	100	$L_0 = 26.2 \text{ mh}$	
	176	?	124	234	$R = 165 \text{ KC/Q}$	
Secondary	50	404	134	58	$C_0 = 6.7 \text{ pf}$	
	100	96	160	97	$L_0 = 24.7 \text{ mh}$	
	180	?	119	235	$R = 155 \text{ KC/Q}$	
$L_p + L_s + 2M$	50	157	137	143	$C_0 = 5.7 \text{ pf}$	Q probably is too low causing R to be too high $L_0$ is correct
	100	35.1	96	406	$L_0 = 62.2 \text{ mh}$	
	114	?	106	419	$R = 390 \text{ KC/Q}$	
$L_p + L_s - 2M$	50	246	109	117	$C_0 = 4.5 \text{ pf}$	
	100	57	103	247	$L_0 = 40.4 \text{ mh}$	
	142	?	75	481	$R = 254 \text{ KC/Q}$	

$M = 5.45 \text{ mh}$ ,  $K = 21\%$

Capacity for 250 KC  
 Primary 15.5 pf  
 Secondary 16.5 pf.

21/3/62

Filter Output Coupler (Obsolete)

Coupler is i.F. transformer of 19/3/62,

Input 0.32 volts from Sig Gen; 52,000 ohm dummy,

11pf primary condenser; H.P. VTVM output meter

Load	56,000 ohms	220,000 ohms	Load Removed			
Freq	Output	Gain	Output	Gain	Output	Gain
KC	Volts	DB	Volts	DB	Volts	DB
150	.05	-10.1	.06	-8.5	.08	-6.0
180	.09	-5.0	.17	0.5	.20	1.9
200	.12	-2.5	.26	4.2	.42	8.4
220	.14	-1.2	.25	3.9	.35	6.8
240	.15	-0.6	.24	3.5	.29	5.2
260	.16	0.0	.24	3.5	.28	4.9
280	.16	0.0	.25	3.9	.30	5.5
300	.16	0.0	.26	4.2	.33	6.3
320	.16	0.0	.27	4.5	.35	6.8
350	.14	-1.2	.24	3.5	.29	5.2
400	.12	-2.5	.17	0.5	.19	1.5
500	.08	-6.0	.10	-4.1	.11	-3.3

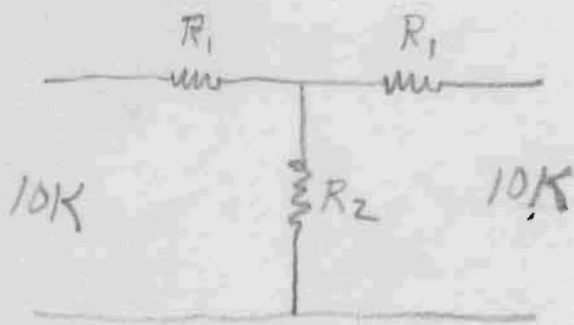
Theoretical

Gain DB      0.3                      6.3                      ?

Primary inductance much too large.

# Filter termination,

22/3/62



$$R_1 = 10K \left( \frac{d-1}{d+1} \right)$$

$$R_2 = 10K \left( \frac{2d}{d^2-1} \right)$$

Filter band, KC	6	12	25	none
(a) Insertion loss DB	4.0	2.7	1.9	0
(b) Bandwidth in DB	0	3.0	6.2	9.2
Required loss in DB	0	4.3	8.3	13.2
$4.0 - a + b$				
Ratio $d$	1.00	1.64	2.60	4.57
$d-1$	0	.64	1.60	3.57
$d+1$	1.00	2.64	3.60	5.57
$d^2$	1.00	2.69	6.76	20.9
$d^2-1$	0	1.69	5.76	19.9
$R_1$ ohms	0	2430	4450	6400
$R_2$ ohms,	$\infty$	19400	9020	4600

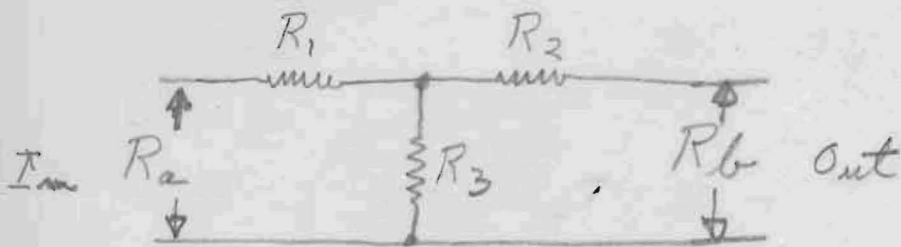
$$d = \frac{\text{load current without attenuator}}{\text{load current with attenuator}}$$

True only for symmetrical networks where input = output = image impedance

# T attenuators

Pender & McIlwain  
1950, p. 11-99

28/3/62



$$d = 2.306 \log_{10} (I_b / I_a)$$

$$R_1 = \frac{(R_a + R_b) \tanh\left(\frac{d}{2}\right) + R_a - R_b}{2}$$

$$R_2 = \frac{(R_a + R_b) \tanh\left(\frac{d}{2}\right) - R_a + R_b}{2}$$

$$R_3 = \frac{(R_a + R_b)}{2 \sinh d}$$

$$\tanh(x) = \frac{e^x - e^{-x}}{e^x + e^{-x}}, \quad \sinh(x) = \frac{e^x - e^{-x}}{2}$$

$$e = 2.718 + \text{ where } x = \frac{d}{2}$$

$$\tanh\left(\frac{d}{2}\right) = \frac{e^d - 1}{e^d + 1}, \quad \sinh\left(\frac{d}{2}\right) = (e^d - 1) / 2e^{d/2}$$



27/4/62

## Filter Output Coupler

Input 0.2 volts from Sig Gen.; 10,000 ohm dummy  
220,000 ohm load; H.P. VTVM output meter

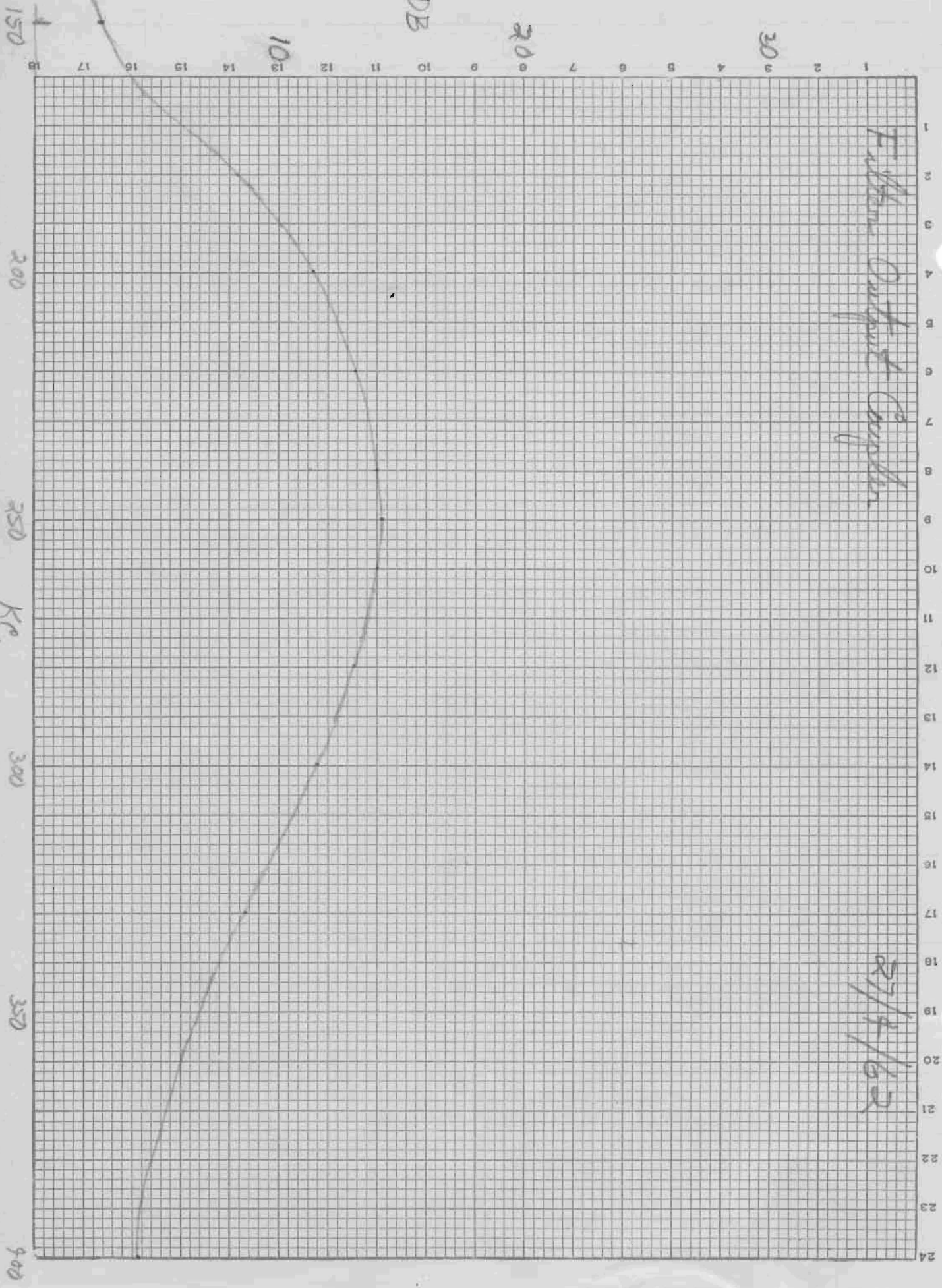
Freq.	Output	Gain
Kc	Volts	DB
150	.14	2.9
180	.26	8.3
200	.37	11.4
220	.45	13.1
240	.50	14.0
250	.51	14.2
260	.50	14.0
280	.45	13.1
300	.38	11.6
330	.27	8.6
400	.16	4.1
500	.10	0.0

Theoretical Gain  $(220,000/10,000)^{1/2} = 4.7 \text{ times} = 13.4 \text{ DB}$   
Circuit slightly undercoupled. Could use a bit larger on  
primary inductance and smaller capacity. This will  
require a marked reduction in primary Co.

(over)

F-105 Output Coupler

27/4/62



30/4/62

1st I.F. with 100K $\Omega$  load, output  $\frac{1}{2}$  volt.

Freq KC	Input mV	Gain DB	Input mV	Gain DB	Input mV	Gain DB
150	40		8.6		7.6	
180	16.2		5.5		5.8	
200	9.5		.90		.92	
205	9.2					
210	9.8		.50		.52	
220	10.7		.50		.52	
230	10.4		.64		.66	
240	11.8		.78		.82	
250	11.8		.78		.80	
260	11.5		.68		.72	
270	10.7		.61		.62	
280	10.1		.52		.54	
290	9.5		.50		.52	
295	9.2					
300	9.6		.66		.67	
330	15.8		2.2		2.4	
400	39		3.0		3.2	
500	71					

Input to low side  
Filter output coupled.  
10000 $\Omega$  dummy.  
Output 2 volts D.C.  
~~1.5 volts D.C.~~  
C

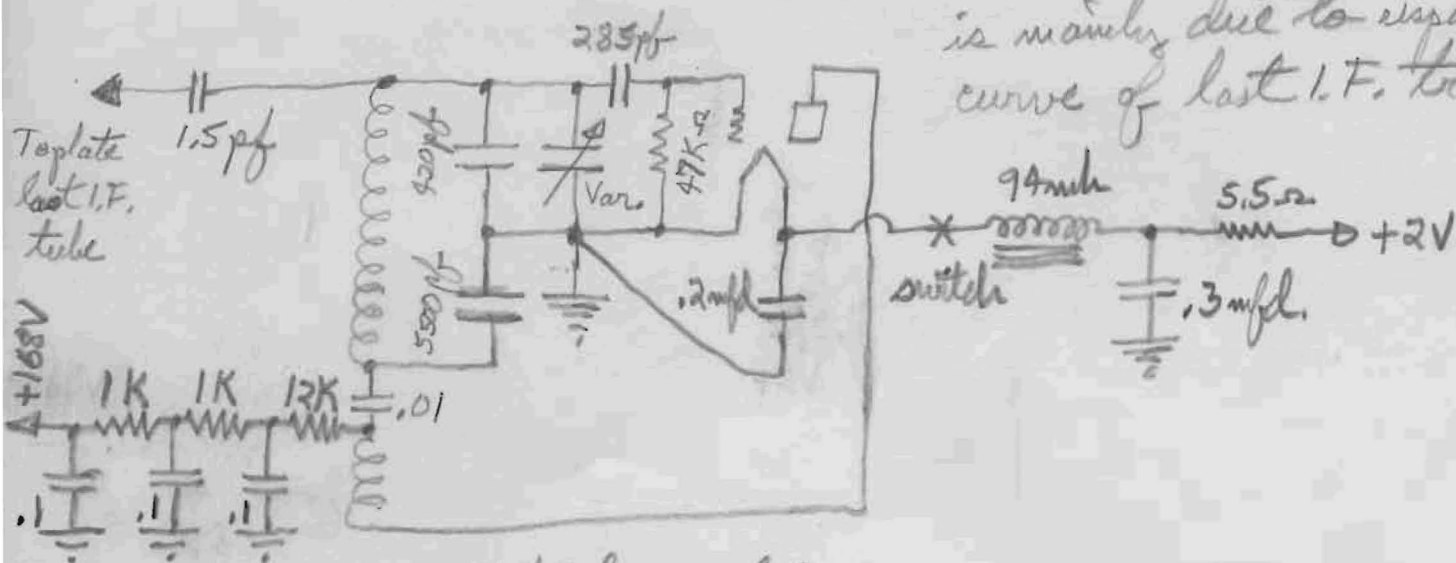
3/5/52

# Beat Oscillator.

Freq KC	Dial T.Tw.	Output Volts	Plate Ma
208	0	1.09	4.08
220	16	1.29	4.05
230	27	1.11	4.02
240	36.5	.97	4.00
250	46	1.00	4.00
260	55	1.11	3.98
270	64.5	1.21	3.97
280	77.5	1.28	3.96
290	94	1.19	3.95
291	100	1.18	3.95

$E_B = 166V$ ,  $E_p = 95V$ , 51.8ma tube.

These readings are taken with last I.F. tube in place and plate voltage turned on. If plate voltage is removed the output volts are slightly (5%) higher due to removal of plate load resistance of last I.F. tube from tuned circuit. The variation of output volts with frequency is mainly due to response curve of last I.F. trans.

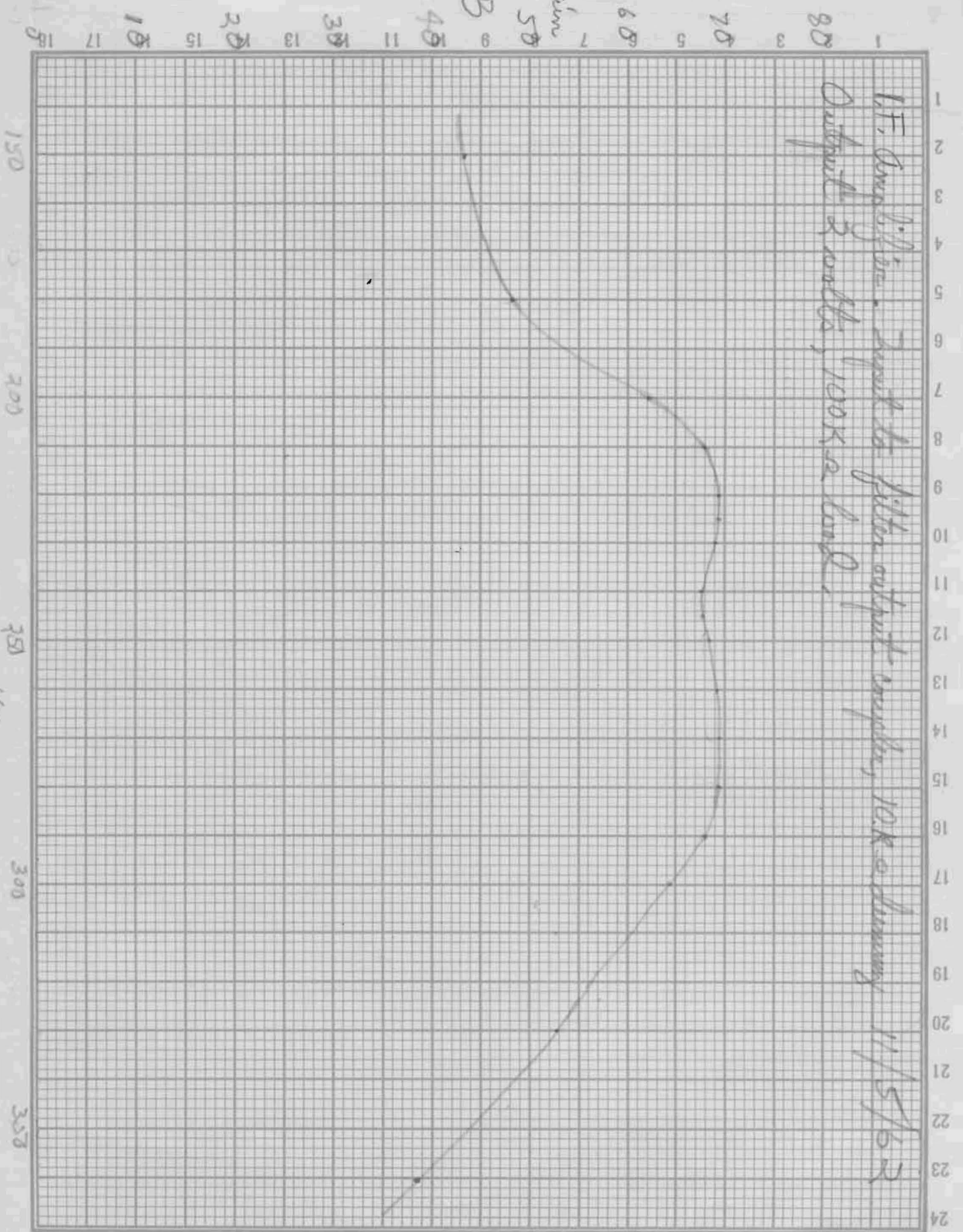


Tube is 1LN5 triode connected.

Variable condenser  $420 + 165 = 585pf$  variation.



LF Amplifier. Input to filter output supply, 10K ohm  
 Output 3 volts, 100K ohm load, 11/5/62



KC

11/5/62

Input to Filter Output Coupler.

10,000 ohm dummy, 56,000  $\Omega$  across 1st I.F. trans.Output 2 volts D.C., 100,000  $\Omega$  load. secondary

KC Freq.	mV input	DB Gain.
-------------	-------------	-------------

150	13.2	43.6
-----	------	------

180	7.8	48.2
-----	-----	------

200	1.52	62.3
-----	------	------

210	0.80	68.0
-----	------	------

220	0.68	69.4
-----	------	------

<del>225</del>	0.68	69.4
----------------	------	------

230	0.70	69.1
-----	------	------

240	0.81	67.8
-----	------	------

<del>246</del>	0.81	67.8
----------------	------	------

250	0.76	68.4
-----	------	------

260	0.71	69.1
-----	------	------

270	0.68	69.4
-----	------	------

280	0.69	69.4
-----	------	------

290	0.80	68.0
-----	------	------

300	1.22	64.3
-----	------	------

330	4.60	52.8
-----	------	------

360	13.6	43.4
-----	------	------

<del>500</del>		
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2/8/67

Mixer Output Performance

Frequency KC	Input mV	Gain DB
150	300	-3.5
180	126	4.0
200	69	10.8
220	45	12.9
230	42	13.6
240	39	14.2
250	38	14.4
260	39	14.2
270	41	13.8
280	43	13.3
300	50	12.0
350	98	6.2
400	172	1.3
500	320	-4.1

Output 0.2V across 10K $\Omega$  load

Input at Mixer grid.

Grid cap removed & 3300 $\Omega$  inserted to prevent mixer grid coil from shorting signal generator.

$$E_p = 160V \quad I_p = 0.6 \text{ ma}$$

$$E_{s_g} = 110V \quad I_{s_g} = 1.0 \text{ ma}$$

$$E_K = 16V \quad I_K = 1.6 \text{ ma}$$

$$E_G = -3.6V$$

$$E_B = 162V \quad E_C = +12V$$

Plate isolation = 3.5K $\Omega$ Screen resistor 50K $\Omega$ Cathode resistor 10K $\Omega$



11/12/62

## Test of Germanium Diodes

Back voltages varied from 0.8 to 1.8 volts.  
 $\Delta E = 1.0$  volts. Type OA95

Back current has considerable scatter. In general  $\Delta I = 0.15$  microamperes when  $I = 1 \mu\text{a}$  at  $E = 1.3\text{V}$  with  $\Delta I$  being proportional to  $I$ . Thus Dynamic back resistance is about

$$\Delta R = 7.5 \cdot 10^6 / I = 7.5 / I \text{ megohms where } I \text{ is in microamperes.}$$

The two diodes placed in receiver have  $I$  about  $0.9 \mu\text{a}$ . Three more diodes with  $I$  between  $1.1$  and  $1.3 \mu\text{a}$  are fastened to inside compartment with adhesive tape.

These are all very temperature sensitive and must be held tightly in pliers when soldering to prevent heat flowing into diode.

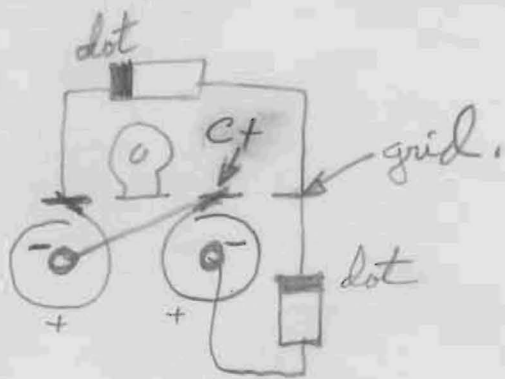
(over)

14/12/62

One of the OA 95 diodes was found to have back current of over  $6 \mu\text{a}$ . Both were removed and replaced by BGE diodes in glass envelopes. These are quite similar and have back currents of about  $4 \mu\text{a}$  each.

also it was found that Mallory cells made poor contact in holder and caused diode to operate at zero voltage. Under this condition the diodes have very low back resistance and act as a short circuit on I.F. transformer.

17/12/61



Diodes installed are Raytheon 1N67.

at 1.35V from Mallory cell.

Back current 3  $\mu$ a

Back resistance 450,000  $\Omega$

Forward current 13 ma

Forward resistance 100 ohms.

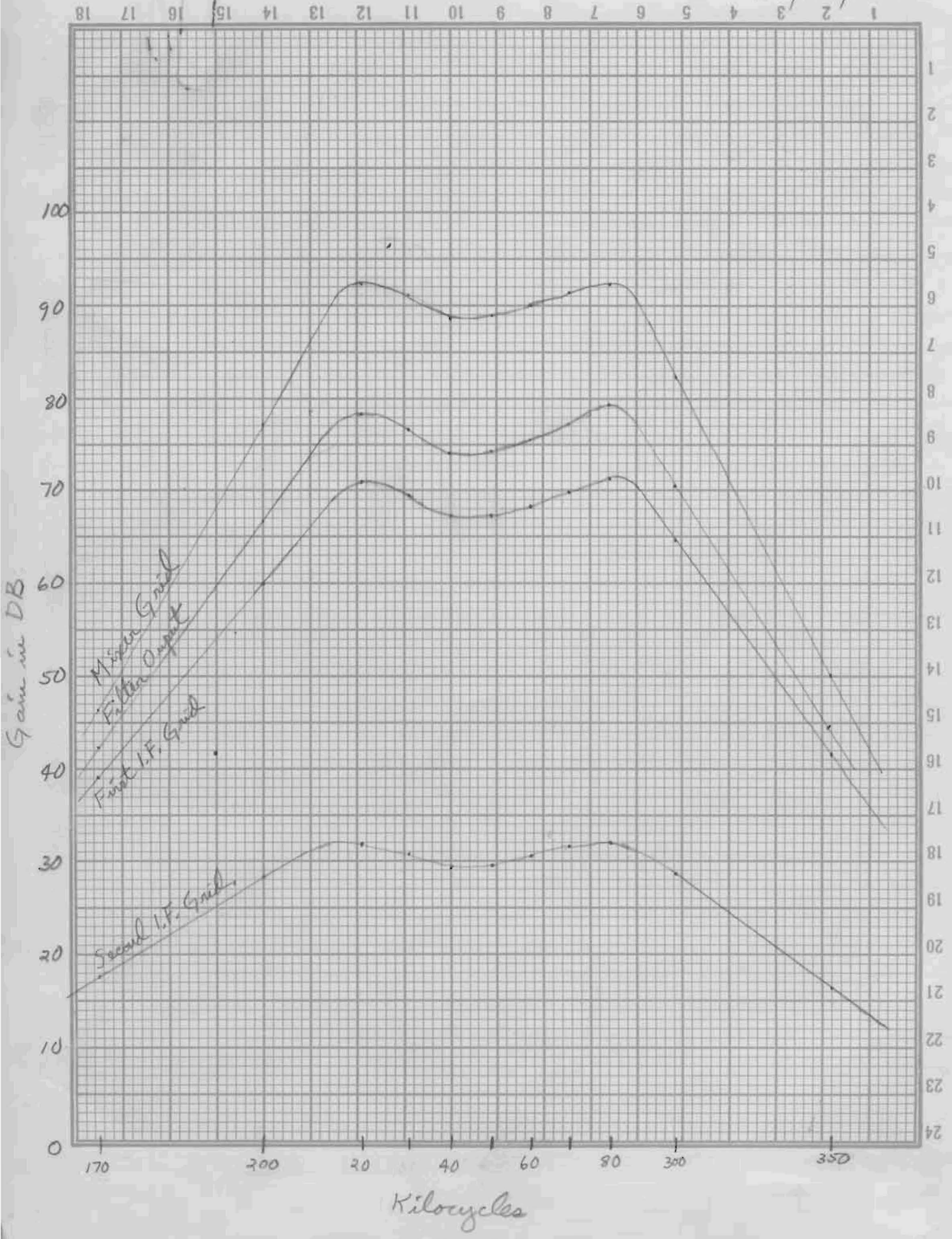
See 19/3/62. for performance of this limiter on output of first I.F. stage. a maximum of 1.53V rms appears across limiter. The waveform will be nearly square. Thus when VTVM indicates 1.53V the peak is actually  $\sqrt{2} \times 1.53 = 2.16$  volts. This is

OK OK top of square wave. Consequently the grid on last I.F. stage should have a negative bias slightly greater than 2.16V, or say 2.5 volts. If this is maintained the grid will never go positive.

11/5/62

One diode had gone bad. Back current 15  $\mu$ a giving a loop resistance of about 100K  $\Omega$ . Replace both diodes with new ones having 1.5  $\mu$ a back current. The stage works satisfactorily again.

14/12/62



## Both I.F. Stages Together

Input 1st I.F. grid. Output 2 volts.  $E_c +12V$

19/3/62

### Gain versus Frequency

Freq. KC	Input mv	Gain DB	Input mv	Gain DB
150	20.0	40.0		
180	11.5	44.8		
200	3.0	56.5		
210	1.78	61.0		
220	222 1.46	1.45 62.7	62.8	
230	1.57	62.1		
240	244 1.78	1.21 61.0	60.9	
250	1.67	61.6		
260	1.56	62.2		
270	278 1.47	1.45 62.7	62.8	
280	1.46	62.7		
290	1.59	62.0		
300	2.2	59.2		
320	5.1	51.9		
350	14.2	43.0		
400				

### Regulation at $E_c = +12V$

$E_B$  Volts 157 166 175

$I_x$  ma 2.05 2.10 2.15

Output % -1.0 100 +1.0



For gain stabilizing versus change in  $B+$   
see 7/1/55 = ~~July 1st 55~~ 7 Jan 55  
on sheets related to 3 stage amplifier.

attempts to reduce gain by operating with very  
small cathode current stop the gain stabilizing  
action.

also see 18/2/57 back of sheet on performance  
of last stage. This agrees with earlier tests,  
apparently a  $15K\Omega$  cathode resistor was used in  
both cases, about  $0.6\text{ ma}$  cathode current  
seems to produce best stability against gain changes  
with changes in Plate to screen volts. This is  
about  $9.0$  volts from cathode to ground.